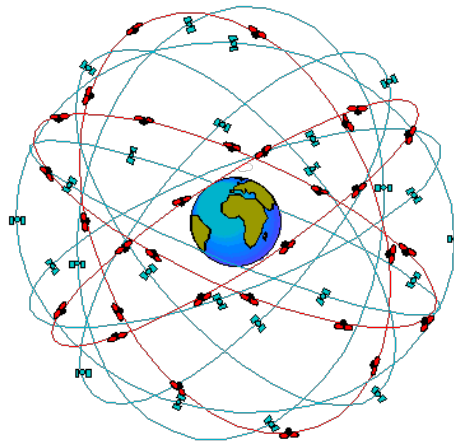


Antenna Phase Center Variation in Multipath Environment



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Precise thinking

Phase Center – where it matters?

- **Ranging applications (positioning, attitude, surveying, etc.)**
- **Imaging applications (radio-telescope, image radars)**
- **Antenna array applications (i.e. null steering)**
- **UWB (signal cohesion and group delay distortion)**



IEEE Standard Phase Centre Definition

The location of a point associated with an antenna such that, if it is taken as the center of a sphere whose radius extends into the far-field, the phase of a given field component over the surface of the **radiation sphere** is “**essentially**” constant, at least over the portion of the surface where the radiation is **significant**.

Notes: Some antennas do not have a unique phase center.

Far-Field condition: $\Delta\psi \leq 22.5^\circ \Rightarrow (\lambda_o/8)$

Source: Antenna Standards Committee of the IEEE
Antennas and Propagation Group



Phase Center Measurements Methods

- **Second Derivative Method [1]**
- **Two Point Method [2]**
- **Edge Diffraction Method [3]**
- **Differential Phase Method [4]**
- **Three Antenna Method [5]**
- **Intuitive for “dummies” Method (this presentation)**
- **Time of Arrival (TOA) Method [6]**



Phase Center Model

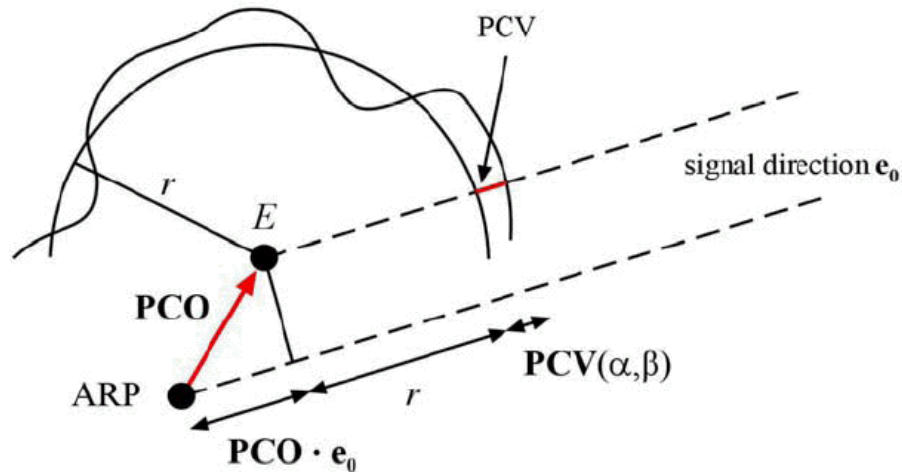


Fig. 1: Antenna model (according to Zeimet and Kuhlmann 2006).

e_0 - is the unit vector of satellite observation
 $obse_0$ - is the unit vector of satellite observation (From sat. to PCO),

r - is the radius of perfect phase sphere originating at E

$$Correction\ Function: \quad S_{ARP} = r + PCO \cdot e_0 + PCV(\alpha, \beta) + \varepsilon \quad (1)$$

PCO is found by minimizing the cost function:

$$\sum (PCV)^2 = Min$$

Phase Center Determination

$$\text{Phase}(\phi, \theta) = \frac{2\pi}{\lambda} (x \cdot \cos \phi \sin \theta + y \cdot \sin \phi \sin \theta + z \cdot \cos \theta) + d$$

or in matrix form:

$$F(\phi, \theta) = \frac{2\pi}{\lambda} [\cos \phi \sin \theta \quad \sin \phi \sin \theta \quad \cos \theta \quad 1] \cdot \begin{bmatrix} x \\ y \\ z \\ d' \end{bmatrix} \Rightarrow F = M \cdot P \quad (1)$$



Phase Center Determination

$$\text{Phase}(\phi, \theta) = \frac{2\pi}{\lambda} (x \cdot \cos \phi \sin \theta + y \cdot \sin \phi \sin \theta + z \cdot \cos \theta) + d$$

or in matrix form :

$$F(\phi, \theta) = \frac{2\pi}{\lambda} \begin{bmatrix} \cos \phi \sin \theta & \sin \phi \sin \theta & \cos \theta & 1 \end{bmatrix} \cdot \begin{bmatrix} x \\ y \\ z \\ d' \end{bmatrix} \Rightarrow F = M \cdot P \quad (1)$$

To find phase center P we use eq. (1), hence :

$$P = M^{-1} \cdot F$$

$$\begin{bmatrix} x \\ y \\ z \\ d' \end{bmatrix} = \frac{\lambda}{2\pi} \begin{bmatrix} \cos \phi \sin \theta & \sin \phi \sin \theta & \cos \theta & 1 \end{bmatrix}^{-1} \cdot F(\phi, \theta)$$



Assemble solution matrix

$$\begin{bmatrix} x^1 & x^2 & : & x^L \\ y^1 & y^2 & : & y^L \\ z^1 & y^2 & : & y^L \\ d^1 & d^2 & : & d^L \end{bmatrix} = \frac{1}{2\pi} \cdot \begin{bmatrix} \cos \phi_1 \sin \theta_1 & \sin \phi_1 \sin \theta_1 & \cos \theta_1 & 1 \\ \cos \phi_2 \sin \theta_1 & \sin \phi_2 \sin \theta_1 & \cos \theta_1 & 1 \\ : & : & : & : \\ \cos \phi_N \sin \theta_1 & \sin \phi_N \sin \theta_1 & \cos \theta_1 & 1 \\ \cos \phi_1 \sin \theta_2 & \sin \phi_1 \sin \theta_2 & \cos \theta_2 & 1 \\ : & : & : & : \\ \cos \phi_N \sin \theta_K & \sin \phi_N \sin \theta_K & \cos \theta_K & 1 \end{bmatrix}^{-1} \cdot \begin{bmatrix} \varphi_1^1(\phi_1, \theta_1) \cdot \lambda^1 & \varphi_1^2(\phi_1, \theta_1) \cdot \lambda^2 & : & \varphi_1^L(\phi_1, \theta_1) \cdot \lambda^L \\ \varphi_2^1(\phi_2, \theta_1) \cdot \lambda^1 & \varphi_2^2(\phi_2, \theta_1) \cdot \lambda^2 & : & \varphi_2^L(\phi_2, \theta_1) \cdot \lambda^L \\ : & : & : & : \\ \varphi_N^1(\phi_N, \theta_1) \cdot \lambda^1 & \varphi_N^2(\phi_N, \theta_1) \cdot \lambda^2 & : & \varphi_N^L(\phi_N, \theta_1) \cdot \lambda^L \\ \varphi_{N+1}^1(\phi_1, \theta_2) \cdot \lambda^1 & \varphi_{N+1}^2(\phi_1, \theta_2) \cdot \lambda^2 & : & \varphi_{N+1}^L(\phi_1, \theta_2) \cdot \lambda^L \\ : & : & : & : \\ \varphi_{N \cdot K}^1(\phi_N, \theta_K) \cdot \lambda^1 & \varphi_{N \cdot K}^2(\phi_N, \theta_K) \cdot \lambda^2 & : & \varphi_{N \cdot K}^L(\phi_N, \theta_K) \cdot \lambda^L \end{bmatrix}$$

where: $\varphi_i = 0 : \frac{2\pi}{N} : 2\pi$, $\theta_i = 0 : \frac{\pi}{2K} : \frac{\pi}{2}$ (upper hemisphere)

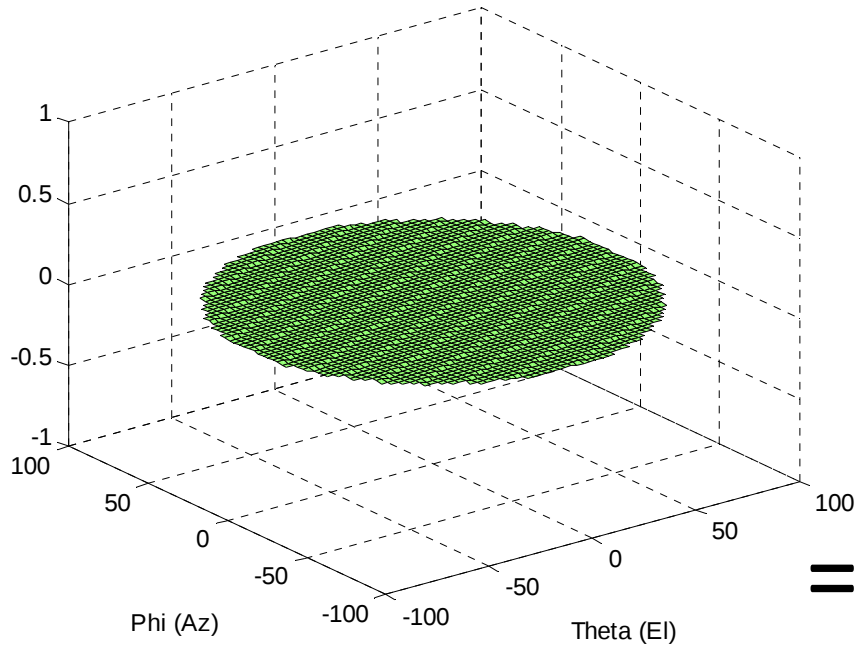
or: $\varphi_i = -\frac{\pi}{2} : \frac{\pi}{N} : \frac{\pi}{2}$, $\theta_i = -\frac{\pi}{2} : \frac{\pi}{K} : \frac{\pi}{2}$ (upper hemisphere)

L : Number of frequencies



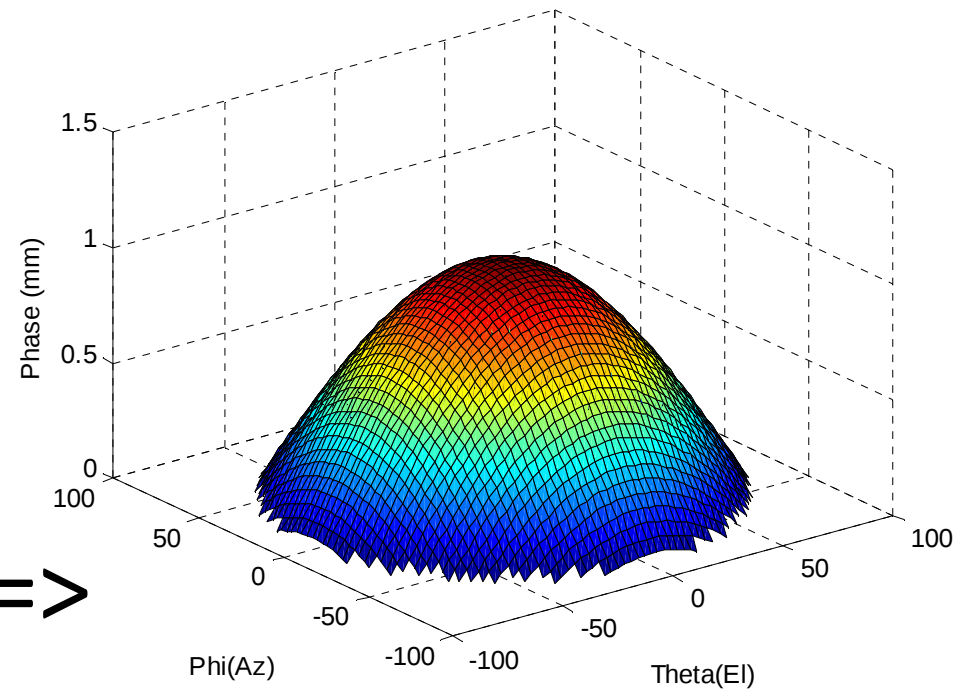
Ideal Phase Patterns

Ideal Phase Pattern at PCO(0,0,0)



$\Rightarrow$

Ideal Phase pattern at PCO(0,0,1mm)



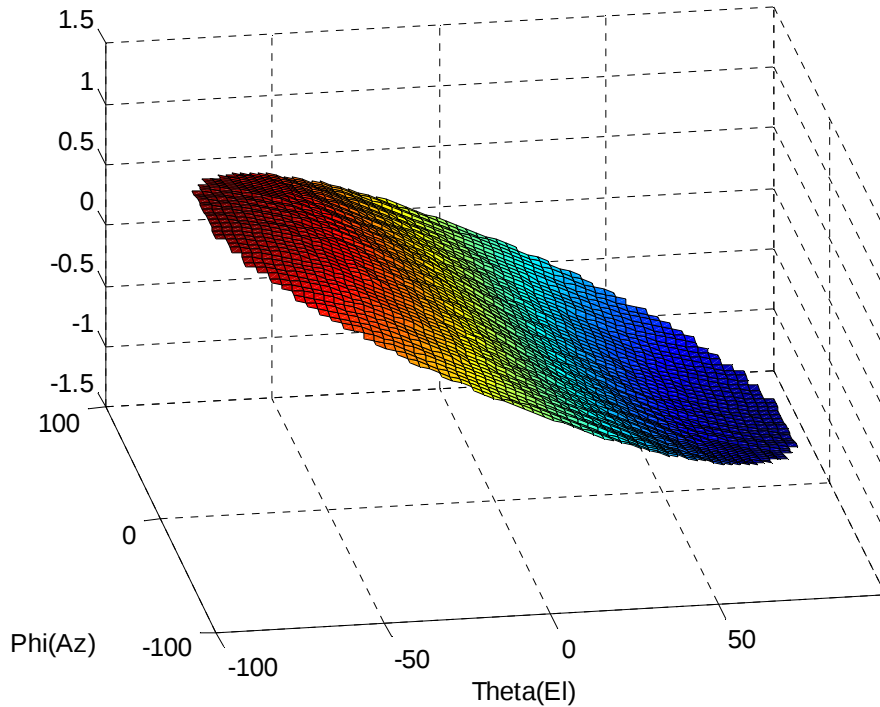
Ideal Phase Pattern shifted along Z-axis from origin (0,0,0) to (0,0,1)



Precise thinking

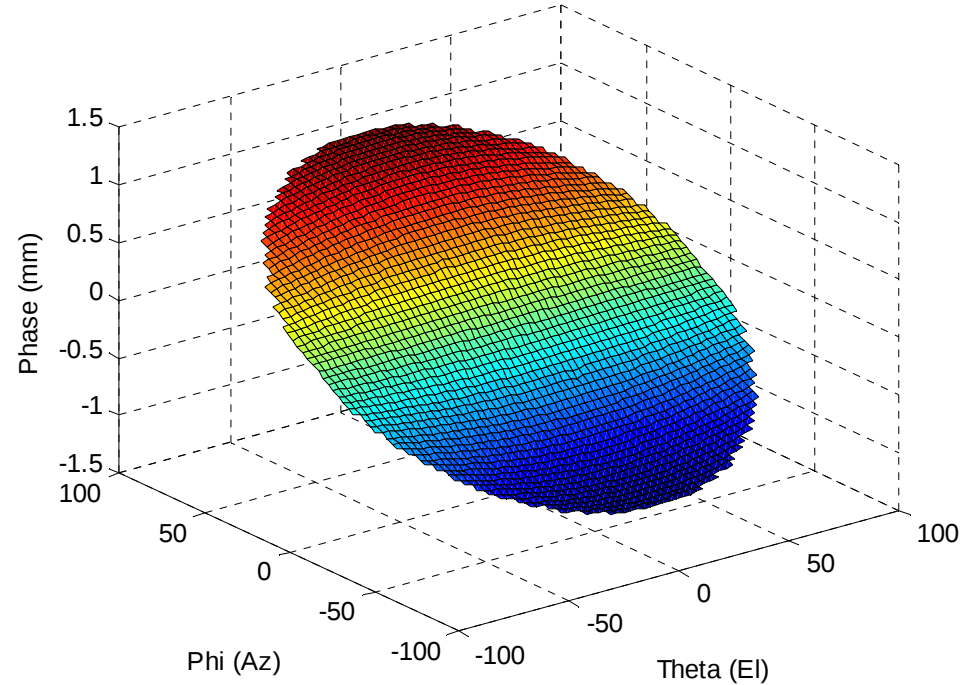
Shifted Ideal Phase Patterns

Ideal Phase pattern at PCO(1,0,0)



Ideal Phase Pattern shifted along
X-axis from origin (0,0,0) to (1,0,0)

Ideal Phase pattern at PCO(0,1,0)

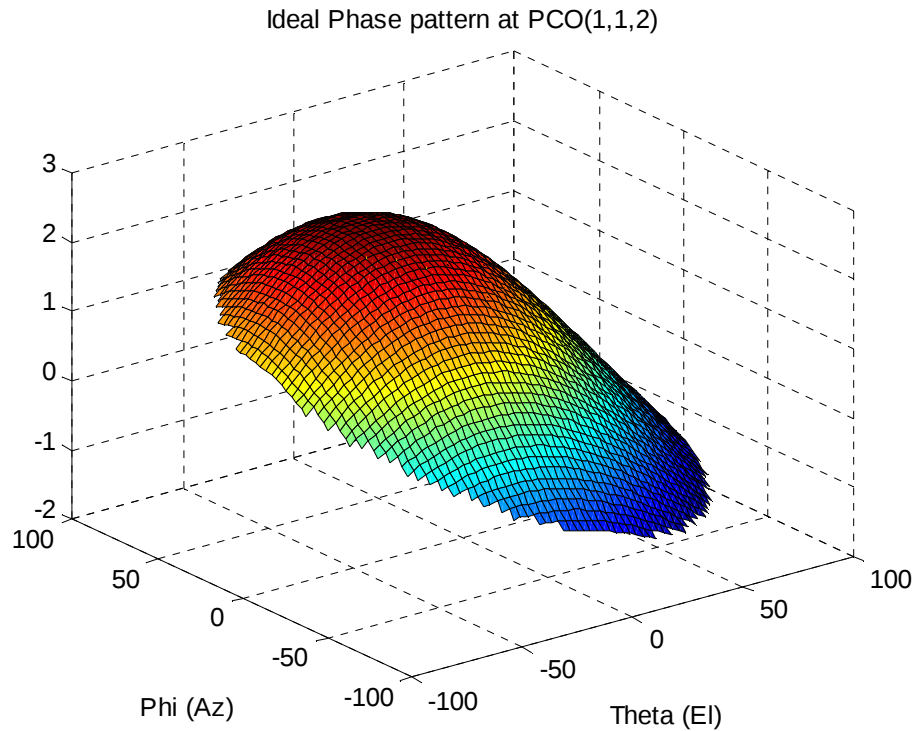


Ideal Phase Pattern shifted along
Y-axis from origin (0,0,0) to (0,1,0)



Precise thinking

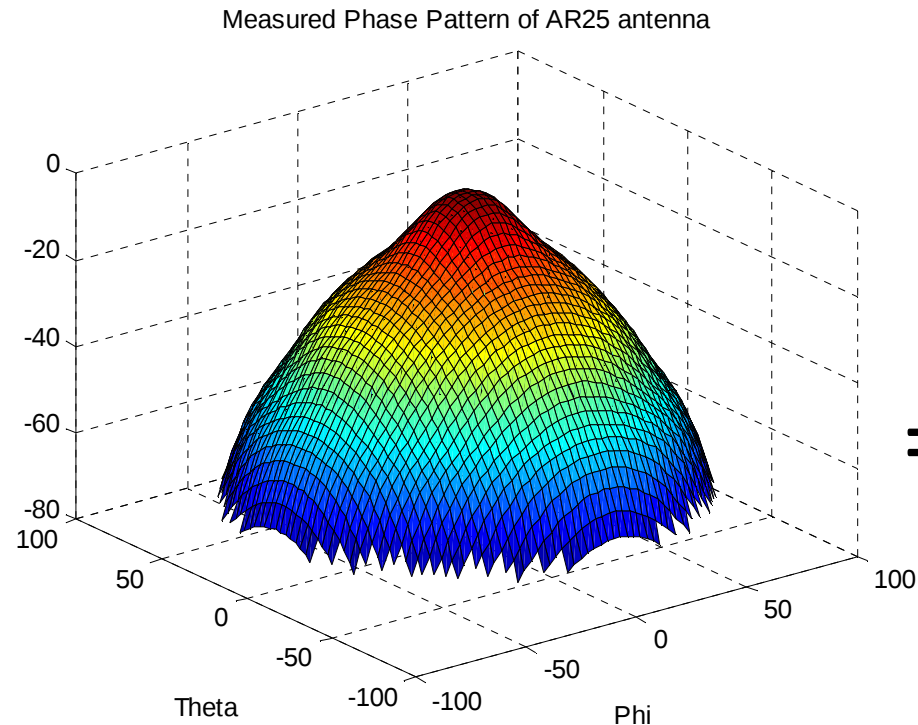
Shifted Ideal Phase Patterns



Ideal Phase Pattern shifted along X,Y,Z-axis from origin (0,0,0) to (1,1,2)



Step 1 – Determine PCO

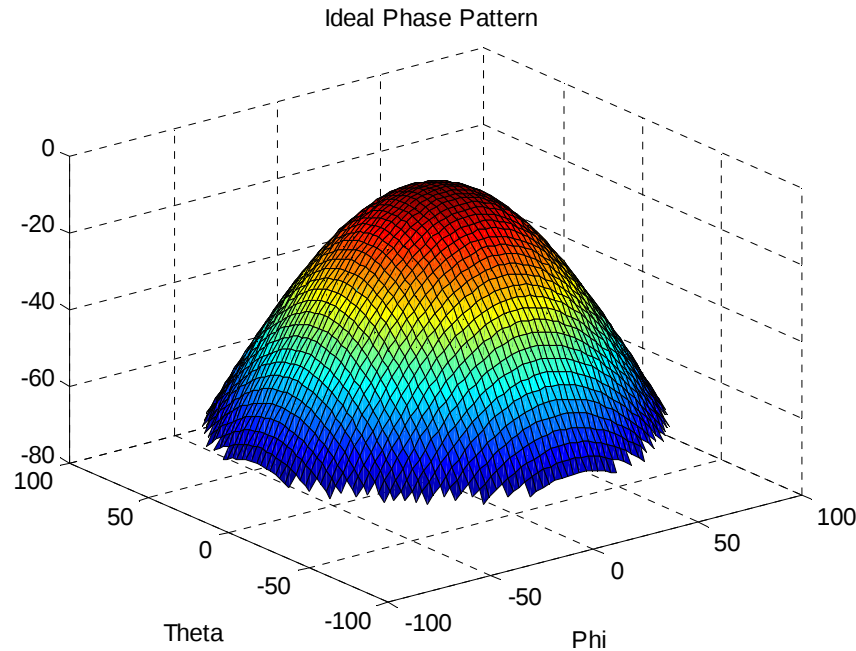


==> Compute mean PCO(x,y,z) using Least Square Method

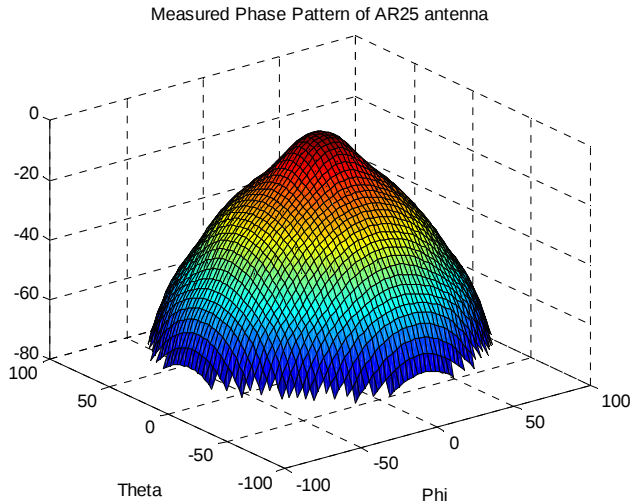
Measured phase pattern of AR25 antenna
(note that this phase pattern shape is heavily weighted by large Z-axis offset of yet to be determined PCO value)



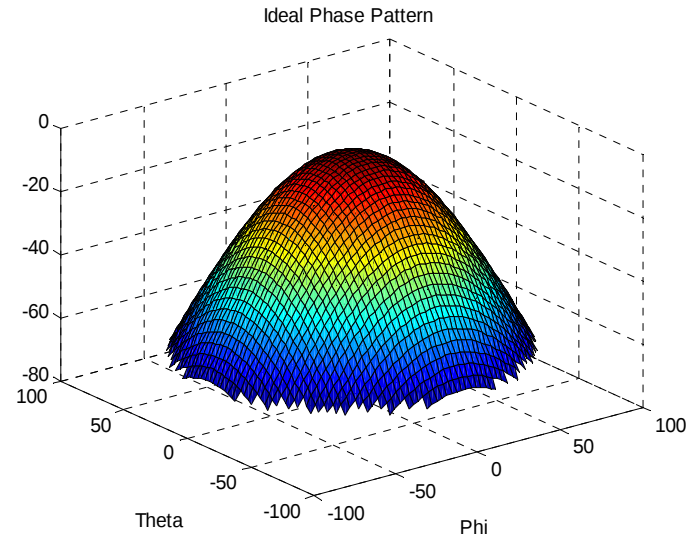
Step 2 – Determine ideal spherical phase pattern at determined PCO location



Step 3 – Determine PCV



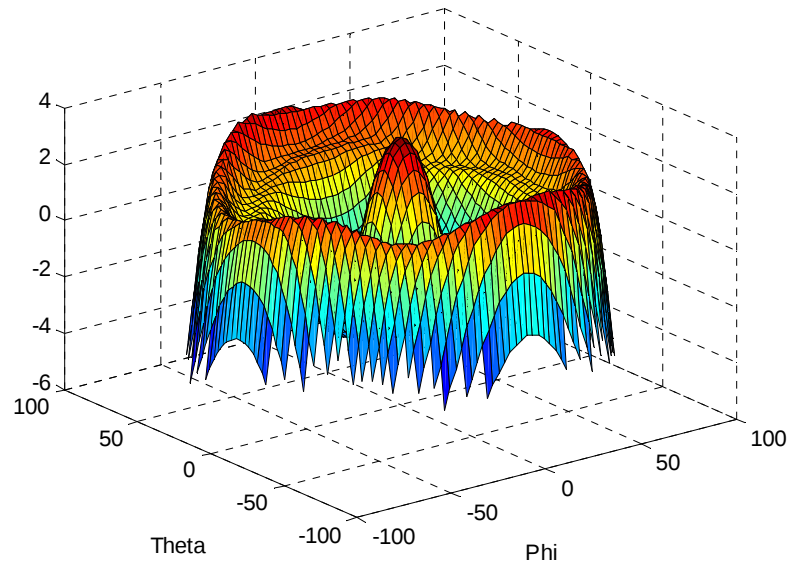
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Computed PCV of AR25 Antenna

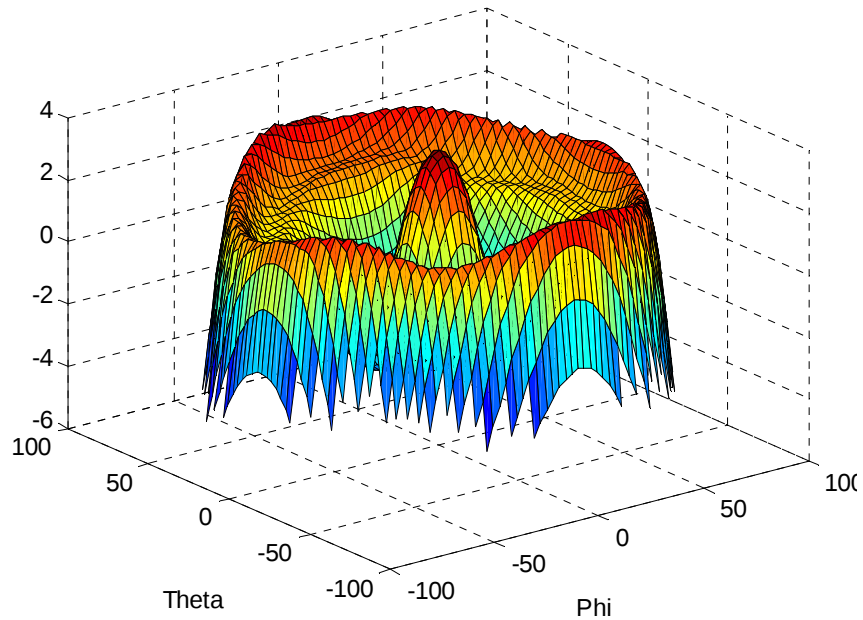
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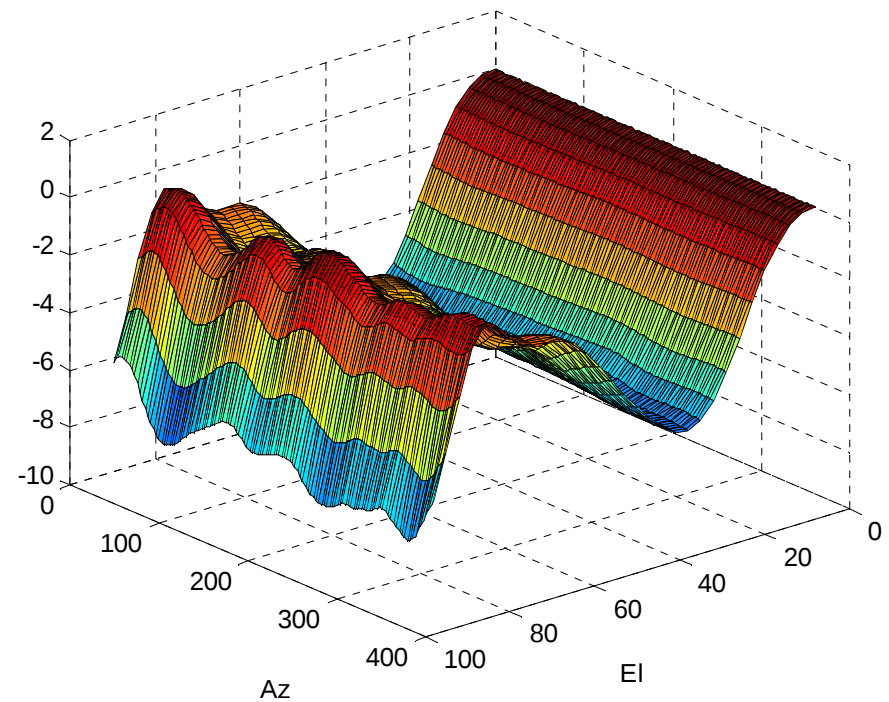
Precise thinking

Computed PCV around PCO:

Computed PCV of AR25 Antenna



Computed PCV of AR25 Antenna

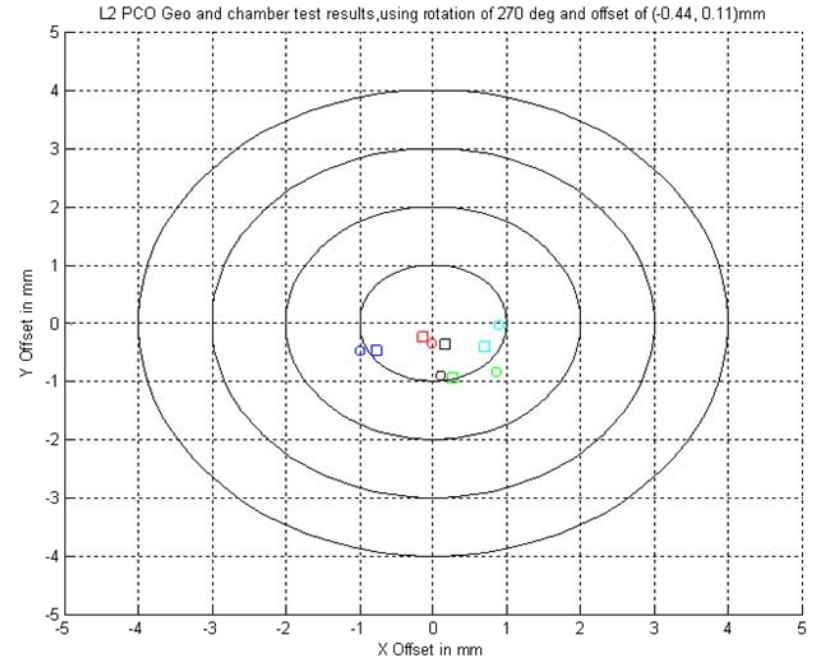
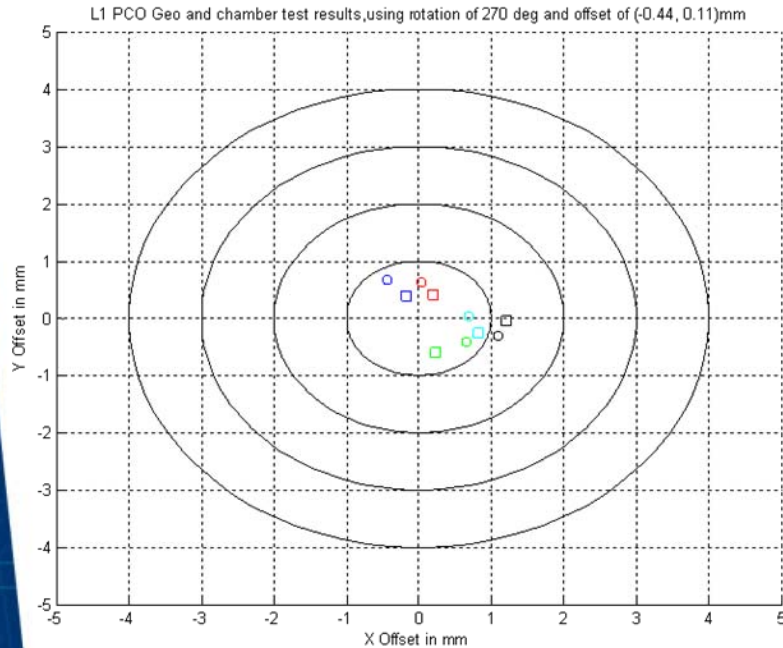


In



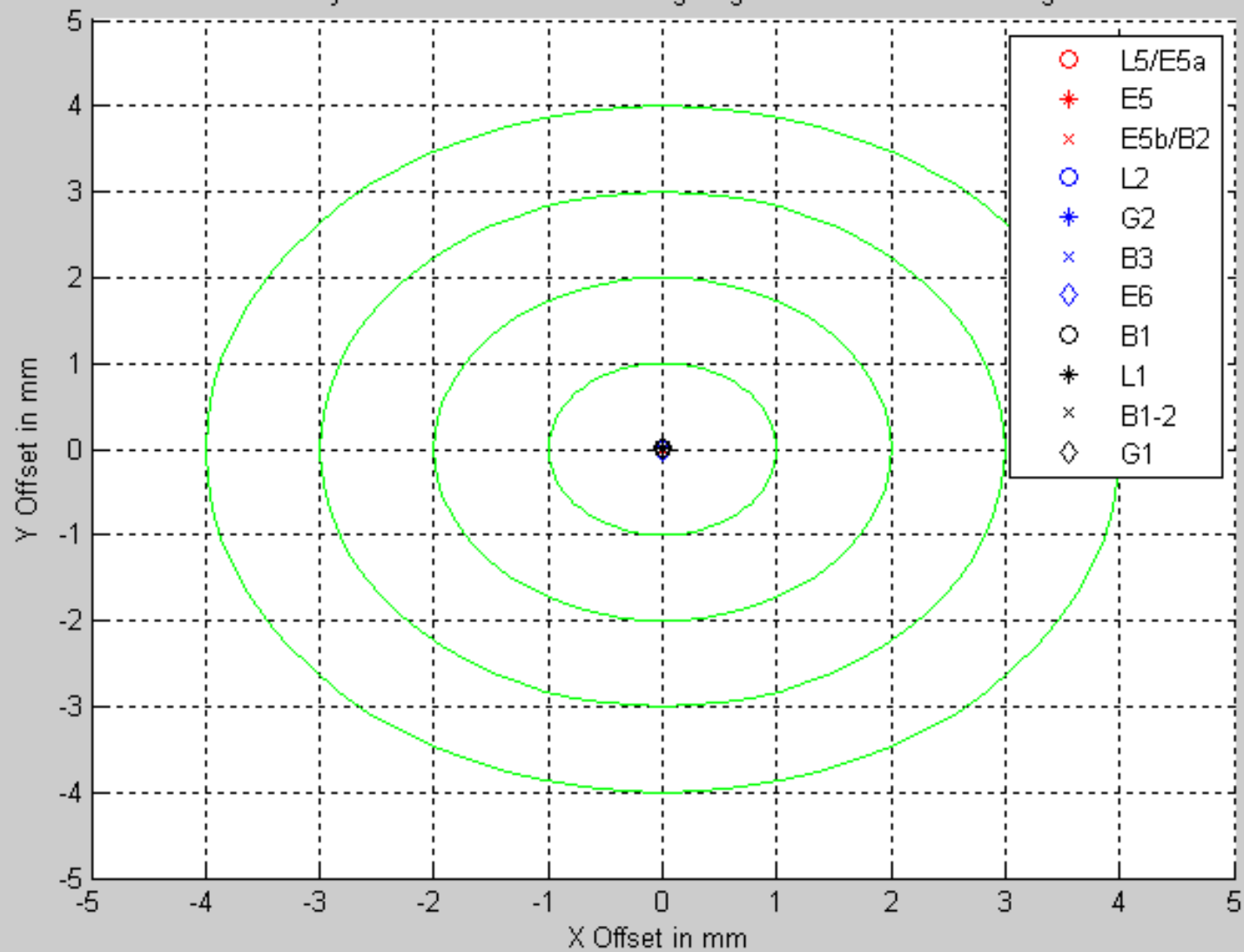
Precise thinking

Example 1 – Anechoic chamber measurements versus Geo++ live GPS signal measurements

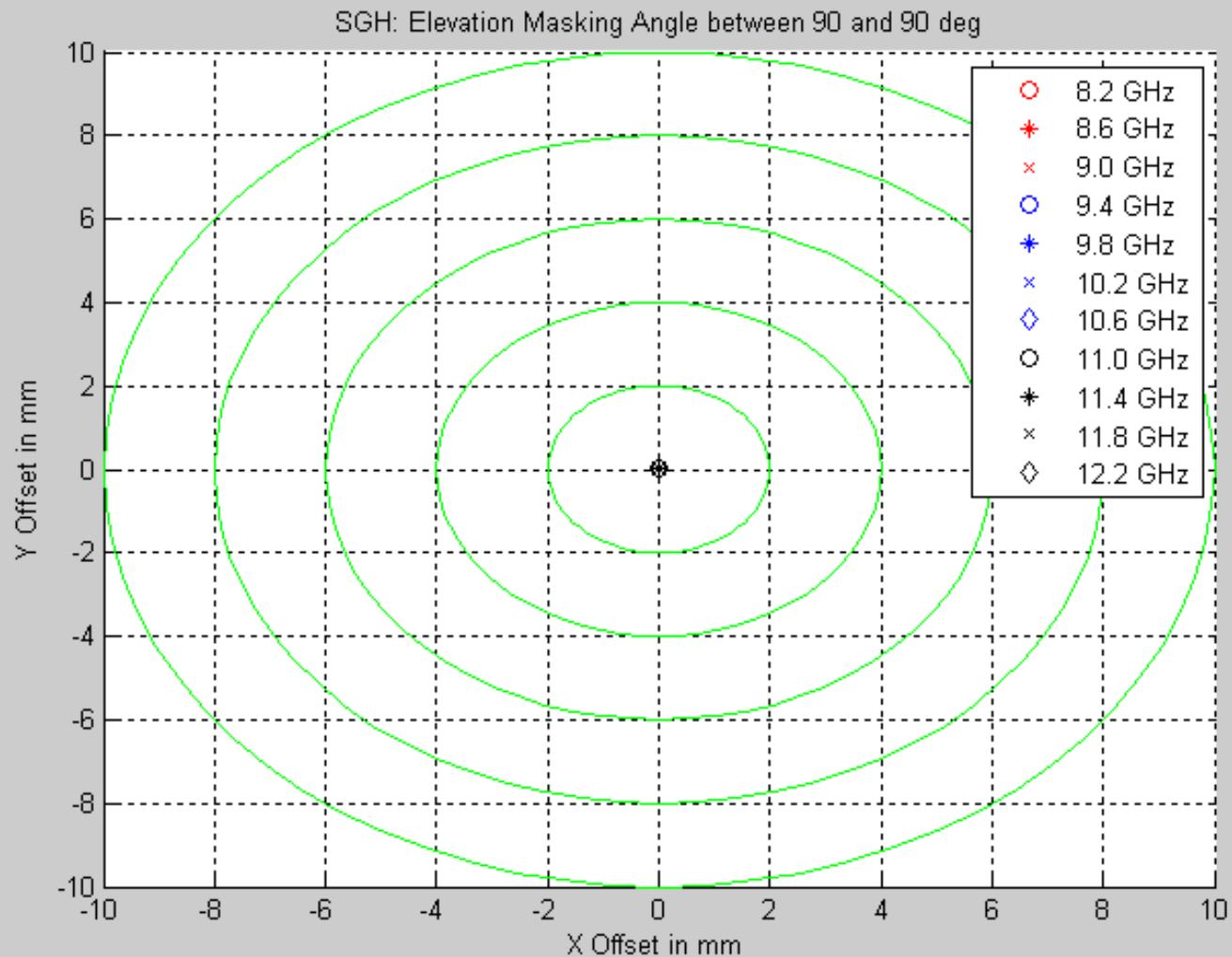


Precise thinking

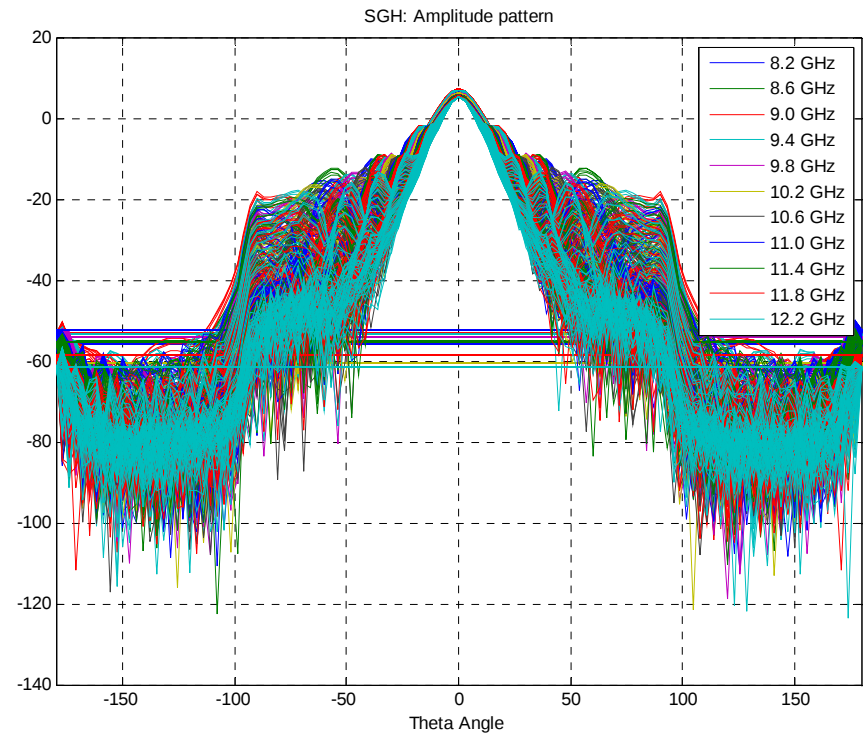
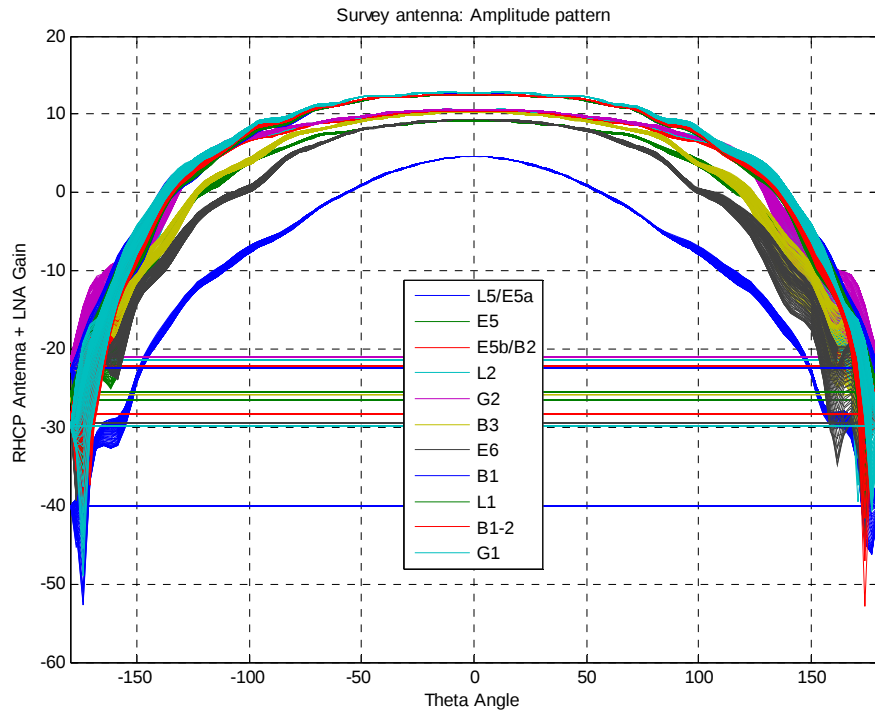
survey antenna: Elevation Masking Angle between 90 and 90 deg



Horizontal Phase Center of SGH Antenna

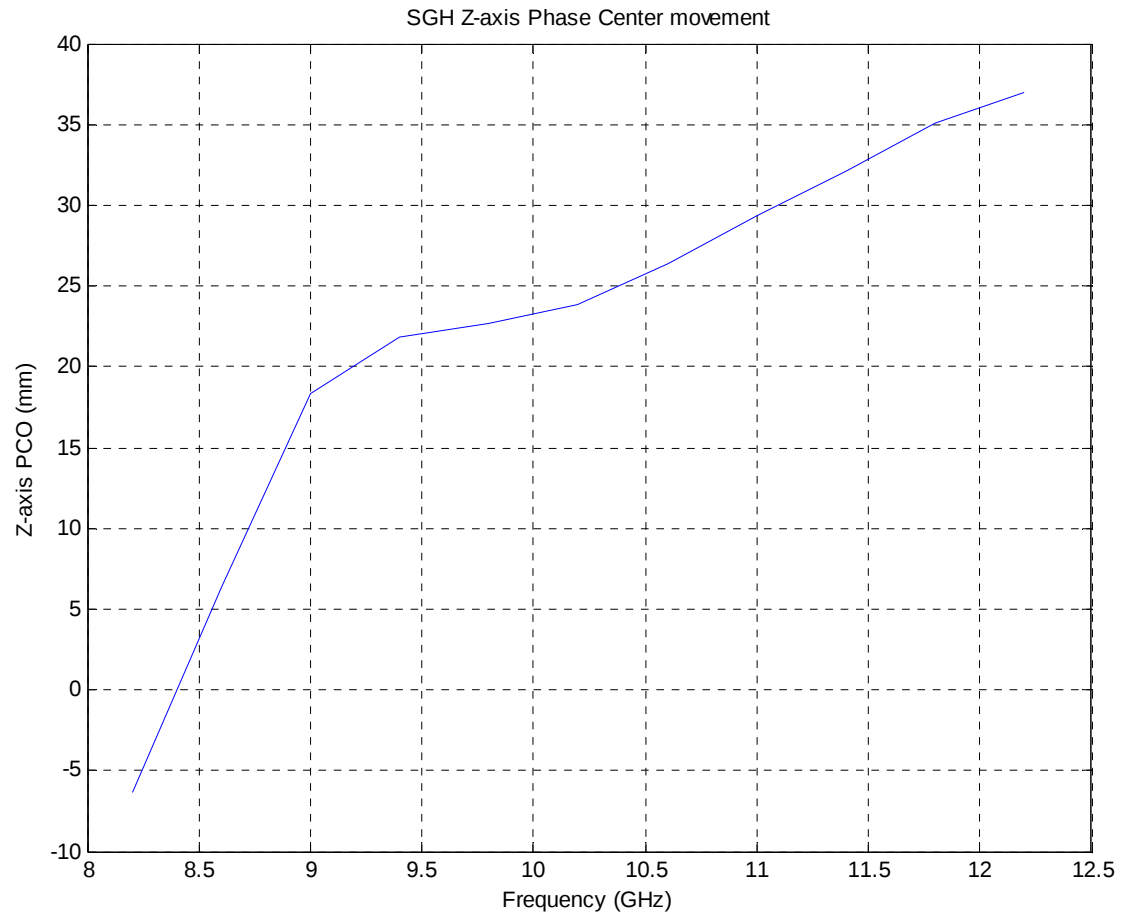
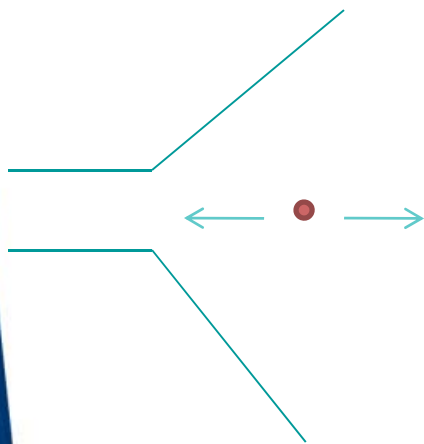


Radiation Amplitude Patterns (GNSS and SGH Antenna)



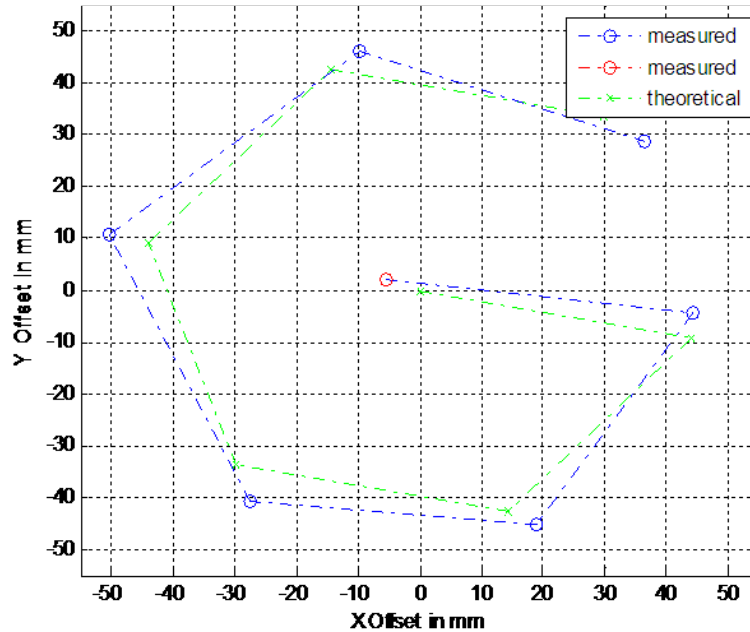
Precise thinking

PCO movement in Z-axis (SGH antenna)

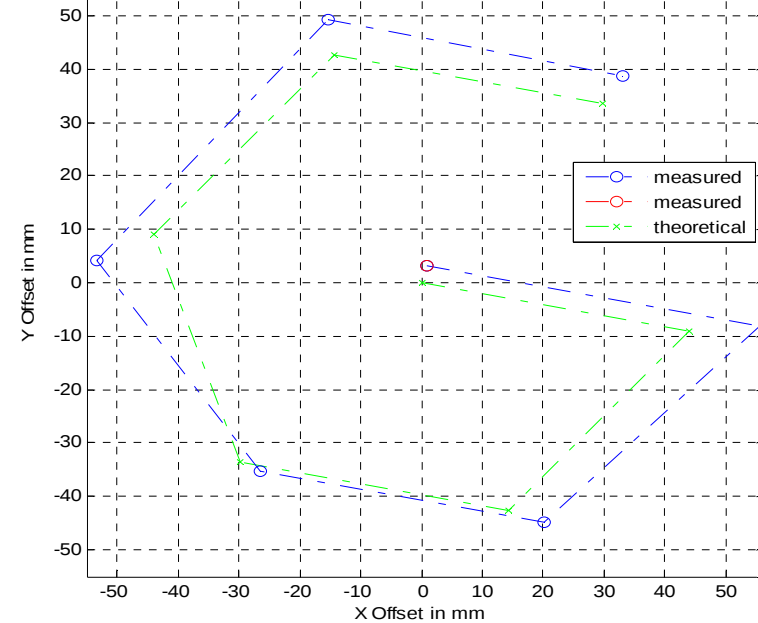


Example 2 - Tightly coupled Array antenna

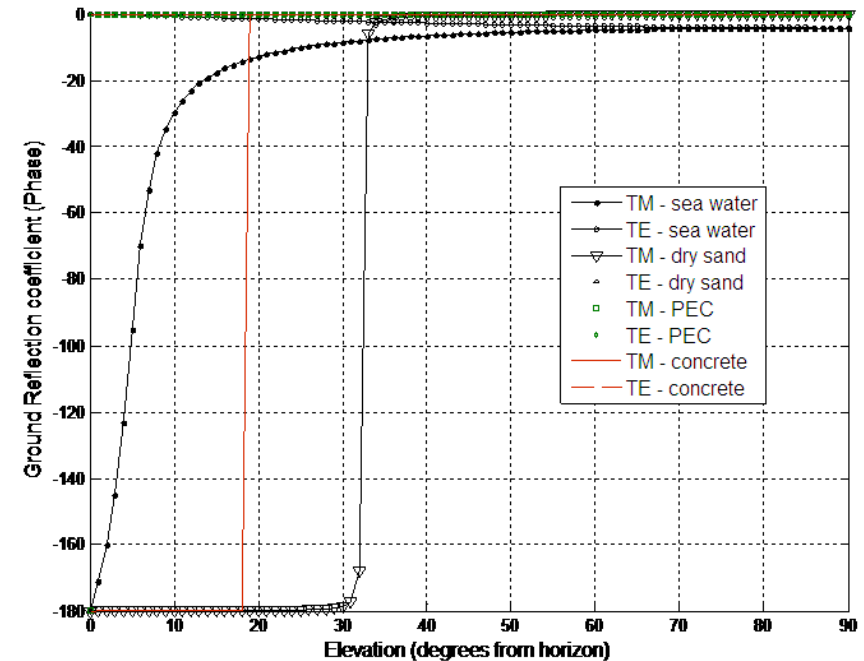
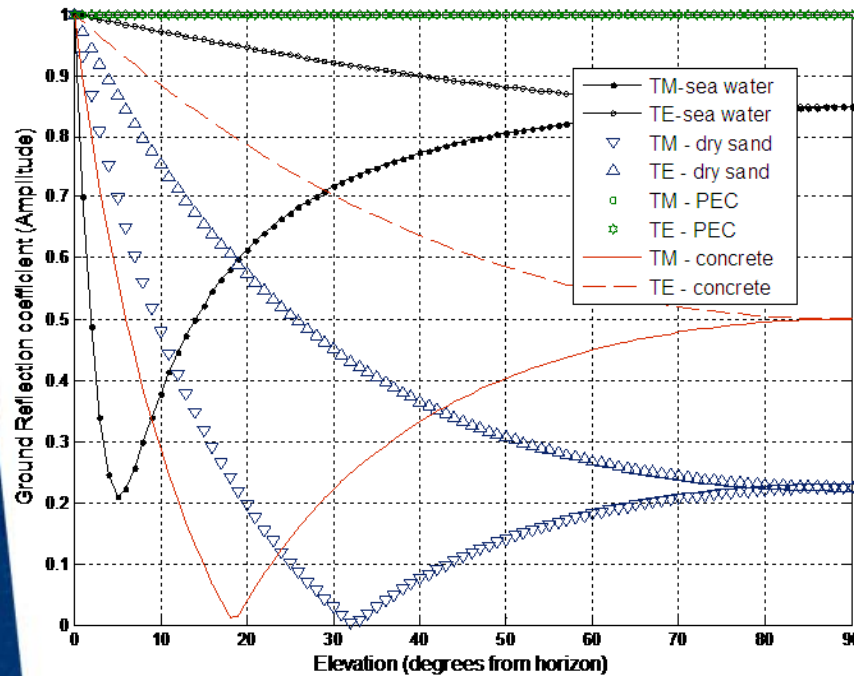
Estimate of electrical positions of the array Rev.G, SN: 01at freq.:1.2276 GHz



Estimate of electrical positions of the array Rev.G, SN: 01at freq.:1.5754 GHz



2-Ray Multipath Model using single reflector



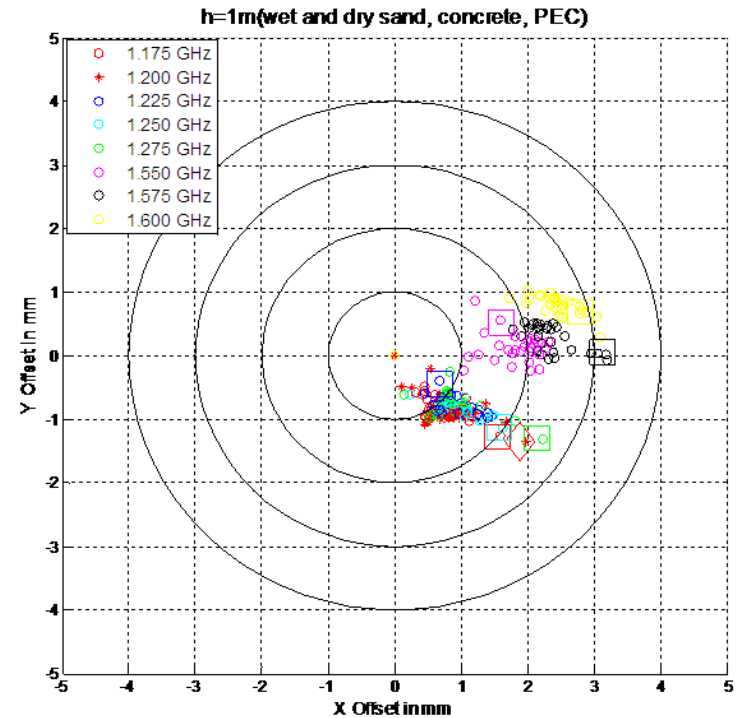
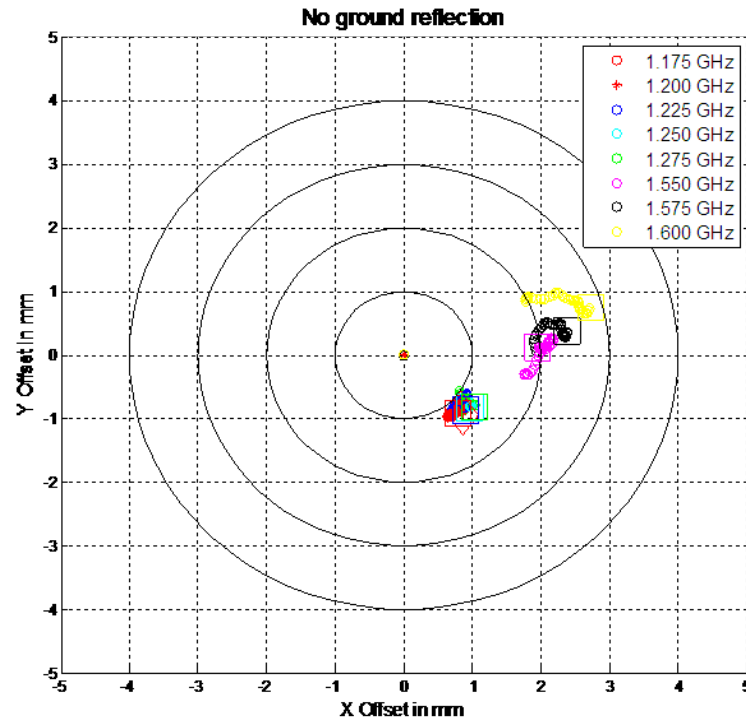
$$\Gamma^{TE} = \frac{\eta_2 \cos \phi^i - \eta_1 \cos \phi^t}{\eta_2 \cos \phi^i + \eta_1 \cos \phi^t}$$

$$\Gamma^{TM} = \frac{\eta_2 \cos \phi^t - \eta_1 \cos \phi^i}{\eta_2 \cos \phi^t + \eta_1 \cos \phi^i}$$



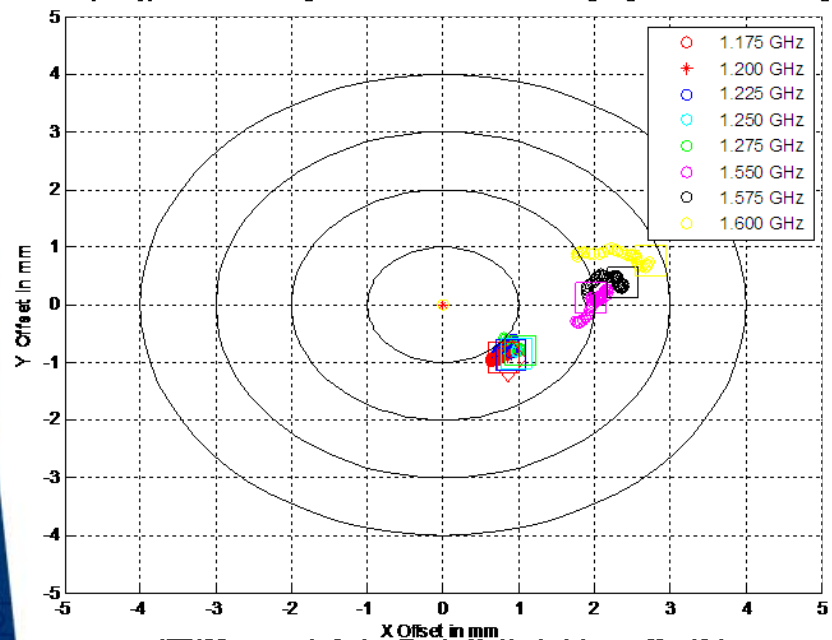
Multipath Effect on PCO

(Example of antenna with good multipath performance)

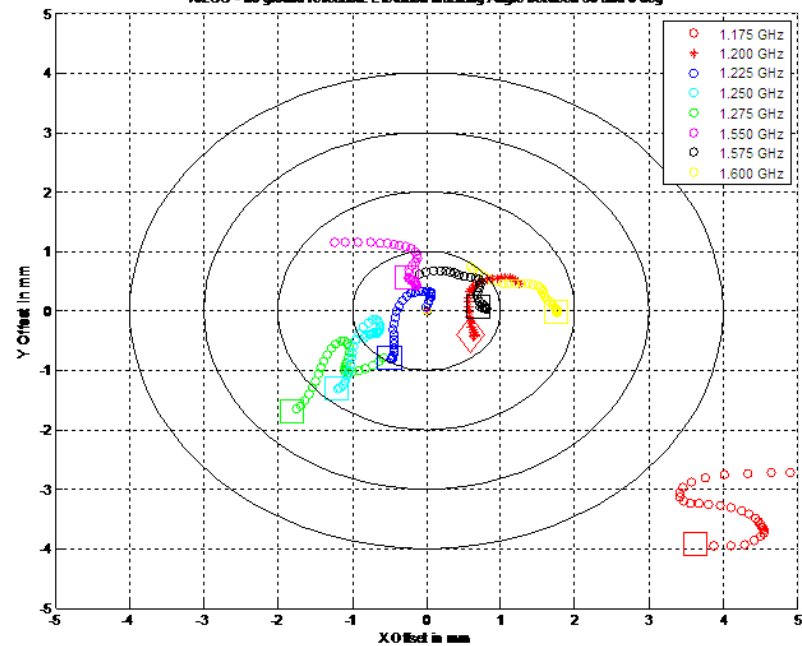


Precise thinking

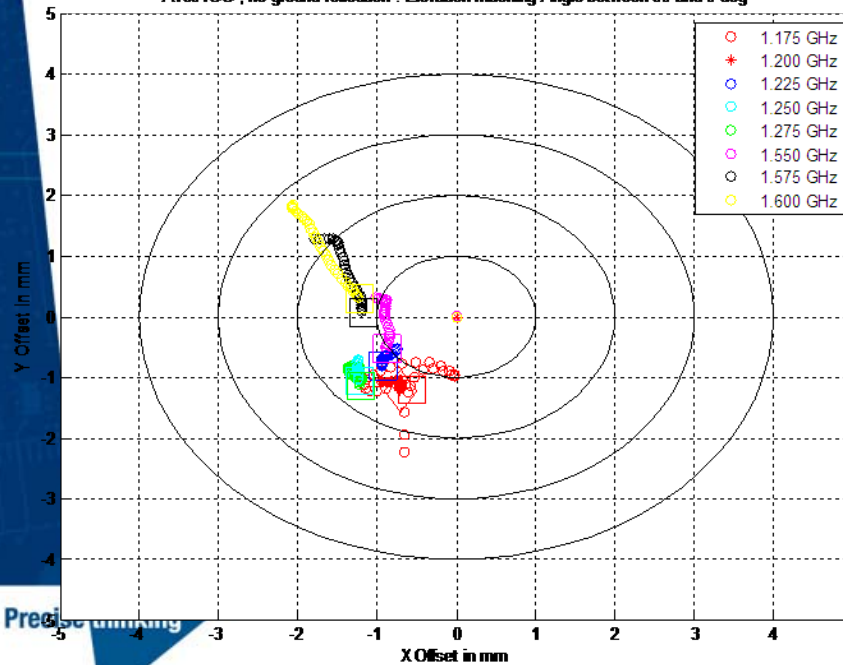
New prototype antenna with no ground reflection : Elevation Masking Angle between 90 and 0 deg



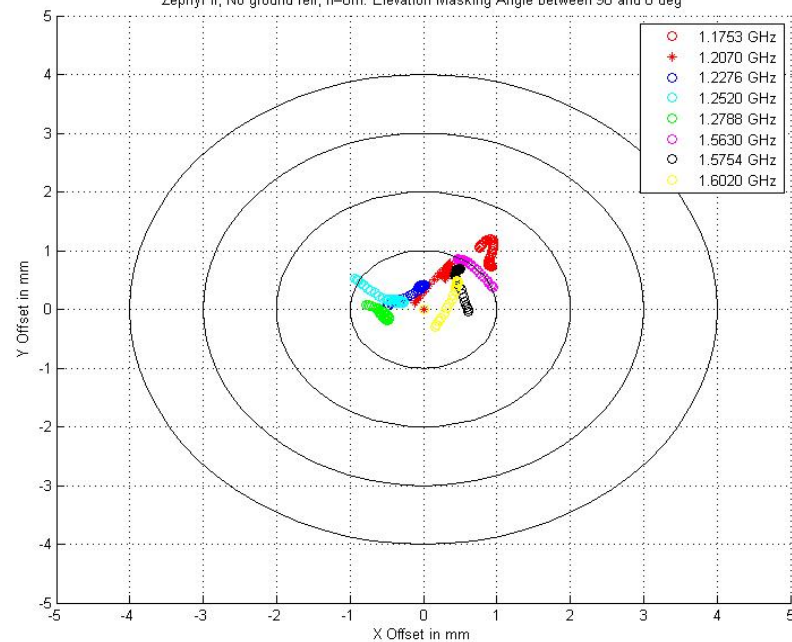
702GG - no ground reflection: Elevation Masking Angle between 90 and 0 deg



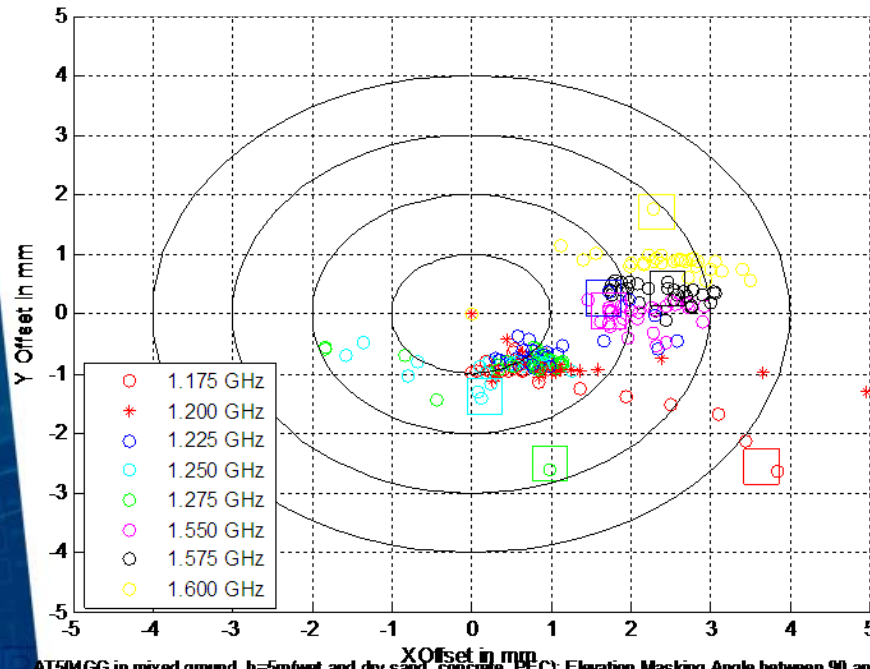
AT514GG , no ground reflection : Elevation Masking Angle between 90 and 0 deg



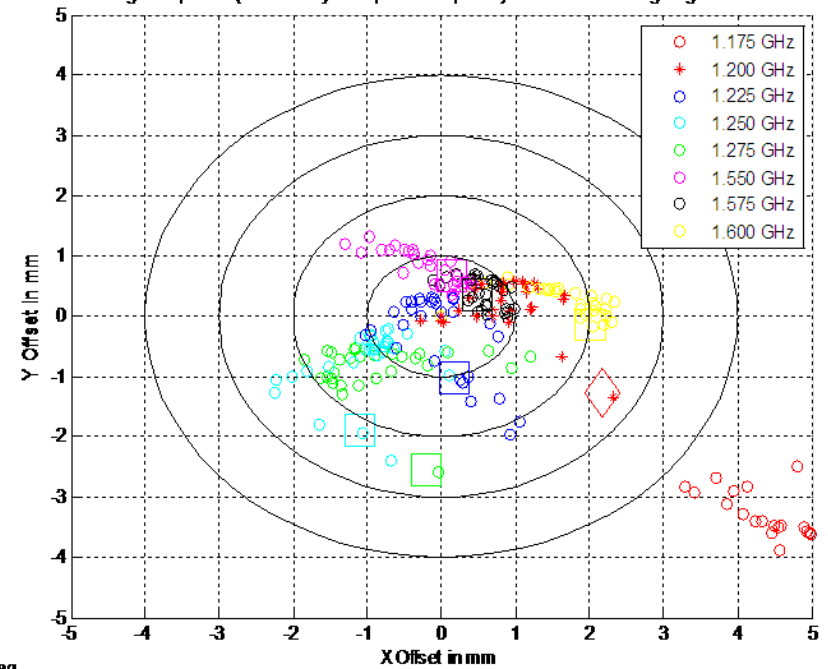
Zephyr II, No ground refl, h=0m: Elevation Masking Angle between 90 and 0 deg



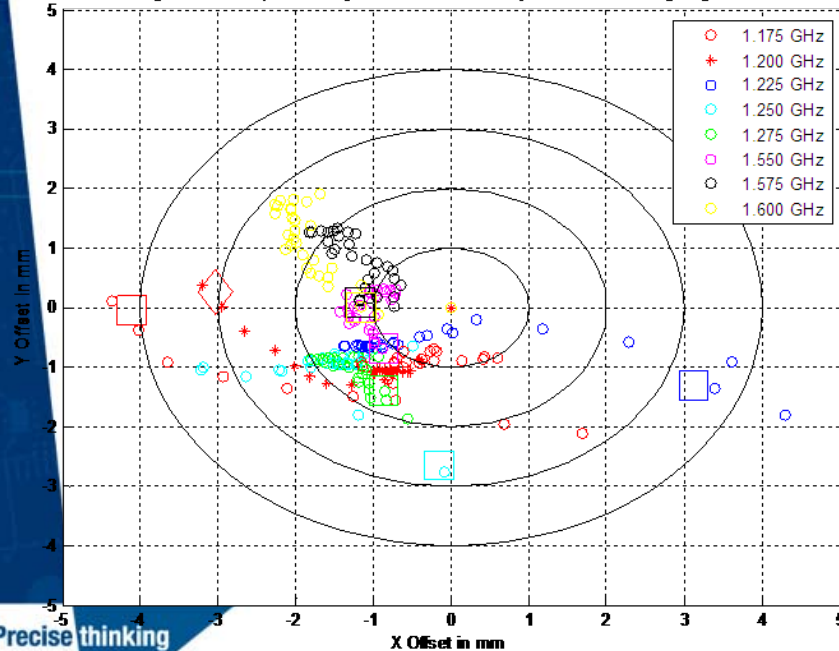
proto antenna in mixed ground, h=5m(wet and dry sand, concrete, PEC): Elevation Masking Angle between 90 and 0 deg



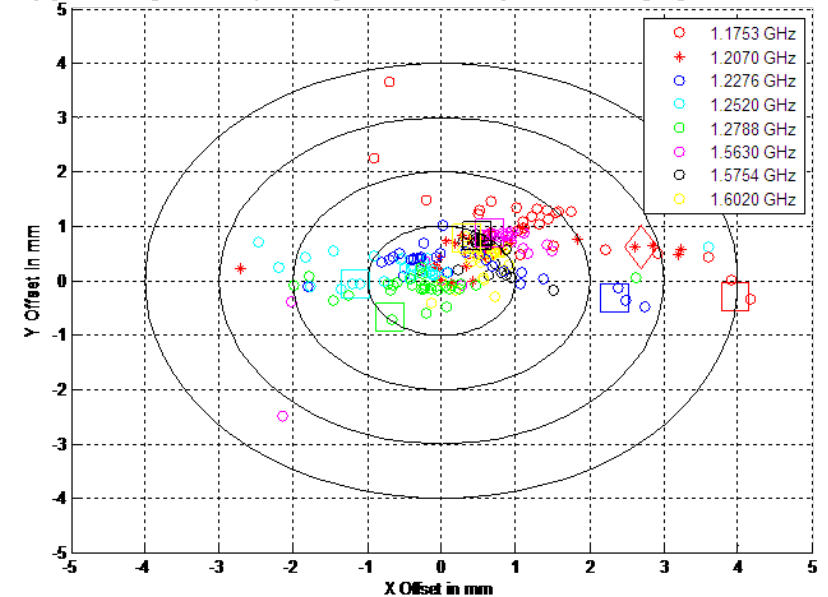
AT506 in mixed ground, h=5m(wet and dry sand, concrete, PEC): Elevation Masking Angle between 90 and 0 deg



AT504GG in mixed ground, h=5m(wet and dry sand, concrete, PEC): Elevation Masking Angle between 90 and 0 deg



Zephyr II in mixed ground, h=5m(wet and dry sand, concrete, PEC): Elevation Masking Angle between 90 and 0 deg



Multipath mitigation

$$\hat{S}(\varphi, \theta) = \hat{C}^*(\varphi, \theta) \cdot \hat{E} + \hat{C}^*(\varphi, -\theta) \cdot \Gamma \cdot \hat{E} \cdot e^{j2\pi d / \lambda}$$

$$\text{where: } \hat{E} = \frac{1}{\sqrt{2}} \begin{pmatrix} 1 \\ -j \end{pmatrix} ; \hat{C}^*(\varphi, \theta) = \frac{a_{\phi, \theta}}{\sqrt{2}} \begin{pmatrix} 1 & j \end{pmatrix} e^{j\varphi(\phi, \theta)}$$

$$\Gamma = \begin{bmatrix} \Gamma^{TE} & 0 \\ 0 & \Gamma^{TM} \end{bmatrix} ; \hat{C}^*(\varphi, -\theta) = \frac{a_{\phi, -\theta}}{\sqrt{2}} \begin{pmatrix} 1 & \pm j \end{pmatrix} e^{j\varphi(\phi, -\theta)} \quad d = 2 \cdot h \sin(\theta)$$

$$\hat{S}^{RHCP} = \frac{a_{R(\theta)} e^{j\delta(\theta)}}{\sqrt{2}} \begin{bmatrix} 1 & j \end{bmatrix} \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & -j \end{bmatrix}$$

$$\hat{S}^{LHCP} = \frac{a_{L(\theta)} e^{j\xi(\theta)}}{\sqrt{2}} \begin{bmatrix} 1 & -j \end{bmatrix} \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & -j \end{bmatrix}$$



Multipath mitigation

$$\hat{S}^{RHCP} = \frac{a_{R(\theta)} e^{j\delta(\theta)}}{\sqrt{2}} [1 \quad j] \frac{1}{\sqrt{2}} [1 \quad -j] + \frac{a_{R(-\theta)} e^{j\delta(-\theta)}}{\sqrt{2}} [1 \quad \pm j] \begin{bmatrix} \Gamma^{TE} & 0 \\ 0 & \Gamma^{TM} \end{bmatrix} \frac{1}{\sqrt{2}} [1 \quad -j] e^{j2\pi d/\lambda}$$

$$\hat{S}^{LHCP} = \frac{a_{L(\theta)} e^{j\xi(\theta)}}{\sqrt{2}} [1 \quad -j] \frac{1}{\sqrt{2}} [1 \quad -j] + \frac{a_{L(-\theta)} e^{j\xi(-\theta)}}{\sqrt{2}} [1 \quad \pm j] \begin{bmatrix} \Gamma^{TE} & 0 \\ 0 & \Gamma^{TM} \end{bmatrix} \frac{1}{\sqrt{2}} [1 \quad -j] e^{j2\pi d/\lambda}$$



Multipath mitigation

$$\hat{S}^{RHCP} = \frac{a_{R(\theta)} e^{j\delta(\theta)}}{\sqrt{2}} [1 \quad j] \frac{1}{\sqrt{2}} [1 \quad -j] + \frac{a_{R(-\theta)} e^{j\delta(-\theta)}}{\sqrt{2}} [1 \quad \pm j] \begin{bmatrix} \Gamma^{TE} & 0 \\ 0 & \Gamma^{TM} \end{bmatrix} \frac{1}{\sqrt{2}} [1 \quad -j] e^{j2\pi d/\lambda}$$

$$\hat{S}^{LHCP} = \frac{a_{L(\theta)} e^{j\xi(\theta)}}{\sqrt{2}} [1 \quad -j] \frac{1}{\sqrt{2}} [1 \quad -j] + \frac{a_{L(-\theta)} e^{j\xi(-\theta)}}{\sqrt{2}} [1 \quad \pm j] \begin{bmatrix} \Gamma^{TE} & 0 \\ 0 & \Gamma^{TM} \end{bmatrix} \frac{1}{\sqrt{2}} [1 \quad -j] e^{j2\pi d/\lambda}$$

$$\hat{S}^{RHCP} = a_{R(\theta)} e^{j\delta(\theta)} + \frac{a_{R(-\theta)} e^{j(2\pi d/\lambda + \delta(-\theta))}}{2} (\Gamma^{TE} \pm \Gamma^{TM})$$

$$\hat{S}^{LHCP} = 0 + \frac{a_{L(-\theta)} e^{j(2\pi d/\lambda + \xi(-\theta))}}{2} (\Gamma^{TE} \pm \Gamma^{TM})$$

$$\hat{S}^{LHCP} = a_{L'(\theta)} e^{j\xi(\theta)} + \frac{a_{L(-\theta)} e^{j(2\pi d/\lambda + \xi(-\theta))}}{2} (\Gamma^{TE} \pm \Gamma^{TM}) \quad \text{where} \quad a_{L'(\theta)} \ll a_{L(\theta)}$$



Multipath mitigation

$$\textit{Scaling Correction}(\phi, \theta) = \frac{a_{R(-\theta)} e^{j[\delta(-\theta) - \xi(-\theta)]}}{a_{L(-\theta)}}$$



Conclusions & Final Remarks

- **A simple method to determine Phase Center Offset/Location based on Least Square Method is described**
- **PCO can be very sensitive to near-by reflections surrounding an antenna**
- **Proper pattern control and controlling amount of x-pol is crucial to keep PCO stable**
- **Multipath mitigation can be employed using information received from well designed dual polarized antenna**



References

- [1] D.Carter, "Phase Centers of microwave antennas", IRE Trans. Antennas and Prop., Vol AP-4, pp 597-600, Oct. 1956
- [2] Y.Y.Hu, "A method of determining phase centers and its applications to electromagnetic horns", J. Franklin Inst., Vol.271, pp 31-39, Jan. 1961
- [3] M. Teichman, "Determination of horn antenna phase centre by edge diffraction theory", IEEE Trans. Aerospace Eelctr. Systems, Vol. AES-9, pp 875-882, Nov. 1973
- [4] J. Thovinen, A. Lehto, A. Raisanenn, "Differential Phase Method", Int. Symp. of Antennas and Propag. Society, Vol. 3, pp 1298-1301, Ma 7-11,1990
- [5] T.W.Hertel, "Phase Center based on the three antenna method", IEEE Int. Symp. of Antennas and Propag. Scoiety, Vol. 3, pp 816-819, June 22-27, 2003
- [6] Y.Yashcheshyn, M.Bury, K.Kurek, P.Bajurko, "Evaluation of the impact of the virtual phase centre effect on the accuracy of the positioning system", 3rd European Conf. on Antennas and Prop., pp 2930-2933, March 23-27, 2009

